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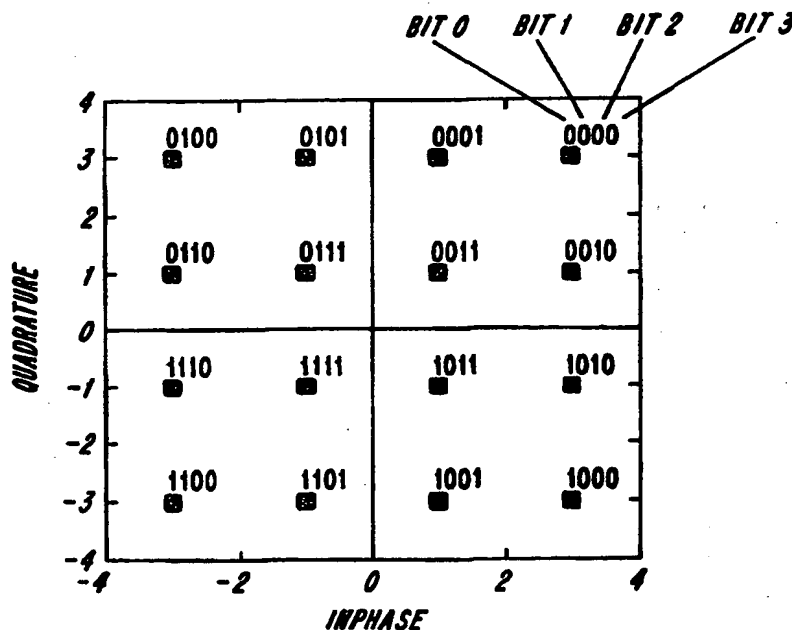
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(54) Title: A METHOD FOR DEMODULATING INFORMATION IN A COMMUNICATION SYSTEM THAT SUPPORTS MULTIPLE MODULATION SCHEMES

(57) Abstract

A method of demodulating voice or data and control information in systems that support multiple modulation schemes modulates voice or data using a first linear modulation scheme, such as 16QAM modulation scheme, and modulates control information using a second linear modulation scheme, for example, QPSK modulation scheme, that has the same symbol rate as that of the first modulation scheme. The first linear modulation scheme has a higher modulation level than the second linear modulation scheme. Information modulated using the second linear modulation scheme, which uses a reduced signal set of the first linear modulation scheme, are demodulated using the same demodulator that is used for demodulating information modulated using the first linear modulation scheme. Also, in-band signalling information within a traffic channel, such as stealing flags, are modulated using the second modulation scheme.



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A METHOD FOR DEMODULATING INFORMATION IN A COMMUNICATION SYSTEM THAT SUPPORTS MULTIPLE MODULATION SCHEMES

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BACKGROUND

This invention generally relates to the field of communication systems and, more particularly, to digital communication systems that supports multiple modulation schemes.

Digital communication systems use a variety of linear and non-linear modulation schemes to communicate voice or data information. These modulation schemes include, Gaussian Minimum Shift Keying (GMSK), Quadrature Phase Shift Keying (QPSK), Quadrature Amplitude Modulation (QAM), etc. GMSK modulation scheme is a non-linear low level modulation (LLM) scheme with a symbol rate that supports a specified user bit rate. In order to increase user bit rate, high-level modulation (HLM) schemes can be used. Linear modulation schemes, such as QAM scheme, may have different level of modulation. For example, 16QAM scheme is used to represent the sixteen variation of 4 bits of data. On the other hand, a QPSK modulation scheme is used to represent the four variations of 2 bits of data. Although 16QAM scheme provides a higher bit rate than QPSK, both of these modulation schemes could have the same symbol rate. Application of modulation schemes, however, differ in many aspects, for example symbol rate and/or burst format, which complicates their support in systems that use multiple modulation schemes.

In wireless digital communication systems, standardized air interfaces specify most of system parameters, including modulation type, burst format, communication protocol, symbol rate, etc. For example, European Telecommunication Standard Institute (ETSI) has specified a Global System for Mobile Communications (GSM) standard that uses time division multiple access (TDMA) to communicate control, voice and data information over radio frequency (RF) physical channels or links using GMSK modulation scheme at a symbol rate of 271 ksps. In the U.S., Telecommunication Industry Association (TIA) has published a number of Interim

Standards, such as IS-54 and IS-136, that define various versions of digital advanced mobile phone service (D-AMPS), a TDMA system that uses a Differential QPSK (DQPSK) modulation scheme for communicating data over RF links.

TDMA systems subdivide the available frequency band into one or
5 several RF channels. The RF channels are divided into a number of physical channels corresponding to time slots in TDMA frames. Logical channels are formed from one or more physical channels, where modulation and channel coding schemes are specified. In these systems, the mobile stations communicate with a plurality of scattered base stations by transmitting and receiving bursts of digital information over uplink and
10 downlink RF channels.

The growing number of mobile stations in use today has generated the need for more voice and data channels within cellular telecommunication systems. As a result, base stations have become more closely spaced, with an increase in interference between mobile stations 12 operating on the same frequency in neighboring or closely
15 spaced cell. Although digital techniques gain more useful channels from a given frequency spectrum, there still remains a need to reduce interference, or more specifically to increase the ratio of the carrier signal strength to interference, (i.e., carrier-to-interference (C/I)) ratio. RF links that can handle lower C/I ratios are considered to be more robust than those that only can handle higher C/I ratios.

20 In order to provide various communication services, a corresponding minimum user bit rate is required. For example, for voice and/or data services, user bit rate corresponds to voice quality and/or data throughput, with a higher user bit rate producing better voice quality and/or higher data throughput. The total user bit rate is determined by a selected combination of techniques for speech coding, channel coding,
25 modulation scheme, and for a TDMA system, the number of assignable time slots per call.

Depending on the modulation scheme used, link quality deteriorates more rapidly as C/I levels decrease. Higher level modulation schemes are more susceptible to low levels of C/I ratio than lower level modulation schemes. If a HLM scheme is
30 used, the data throughput or grade of service drops very rapidly with a drop in link

quality. On the other hand, if a LLM scheme is used, data throughput or grade of service does not drop as rapidly under the same interference conditions. Therefore, link adaptation methods, which provide the ability to change modulation and/or coding based on the channel conditions, are used to balance the user bit rate against link quality.

- 5 Generally, these methods dynamically adapt a system's combination of speech coding, channel coding, modulation, and number of assignable time slots to achieve optimum performance over a broad range of C/I conditions.

One evolutionary path for the next generation of cellular systems is to use high-level modulation (HLM), e.g., 16QAM modulation scheme or 8PSK, to provide
10 increased user bit rates compared to the existing standards. These cellular systems include enhanced GSM systems, enhanced D-AMPS systems, International Mobile Telecommunication 2000 (IMT-2000), etc. A high level linear modulation, such as 16QAM modulation scheme, has the potential to be more spectrum efficient than, for example, GMSK, which is a low-level modulation (LLM) scheme. Furthermore, the
15 use of 16QAM modulation scheme in conjunction with a higher symbol rate significantly increase the user bit rate compared to the GMSK modulation scheme. In this way, the maximum user bit rate offered by an HLM scheme, such as 16QAM modulation scheme, may be more than doubled. Because higher level modulation schemes require a higher minimum C/I ratio for acceptable performance, their availability in the system
20 becomes limited to certain coverage areas of the system or certain parts of the cells, where more robust links can be maintained. However, a system can be planned to provide full coverage for HLM scheme. The modulation schemes provided in a cell may be a mixture of non-linear and linear modulation, with different symbol rates.

Generally, two types of logical channels are defined by air interface
25 standards: control channels (CCH) and traffic channels (TCH). CCHs are used for control signalling such as registration, authentication, call set-up, and the like. TCHs, which are single user channels, are used to handle voice or data communication. For TCHs, some of the standards define various user bit rates.

In GSM systems, control signalling is carried out using different types of
30 CCHs, including dedicated control channels (DCCHs), Broadcast Channels (BCHs),

and Common Control Channels (CCCHs). BCHs include Frequency Correction channel (FCCH), Synchronization Channel (SCH), and Broadcast Control Channel (BCCH). The CCCHs include Paging channel (PCH), Access Grant Channel (AGCH) and Random Access Channel (RACH). DCCHs include Stand-alone Dedicated Control Channel (SDCCH), Fast Associated Control Channel (FACCH), and Slow Associated Control Channels (SACCH).

FCCH indicates a BCCH carrier signal and enables a mobile station to synchronize to its frequency. SCH is used to signal TDMA frame structure in a cell and a Base Station Identity Code (BSIC) that indicates whether a base station belongs to a GSM system or not. BCCHs is transmitted during predefined time slot (e.g., time slot 0 in single carrier base stations) of a downlink RF channel, to provide general information to the mobile stations. SDCCH, which may be transmitted at a time slot adjacent to BCCH, is used for registration, location updating, authentication and call set-up. PCH is a downlink only channel, which is used for informing the mobile station of a network's signaling requirement, for example when the mobile unit is called. AGCH is a downlink only channel used for replies to access requests for assigning a dedicated control channel for a subsequent signaling. RACH is used by a mobile station to request a channel, when it is paged, or when it wants to initiate a call.

The associated control channels, FACCH and SAACH are always associated with traffic channels. Applicable standards specify a number of bits for FACCH and SACCH, which are communicated according to a pre-defined format. SACCH is used for communicating control and supervisory signals associated with traffic channels, including the transmission of parameters corresponding to a measure of bit error rate (BER) or a measure of received signal strength (RSS) at mobile stations.

12. FACCH steals bursts allocated for traffic channels for control requirements, such as hand-over.

Fast signaling procedures are needed to quickly provide signalling information to the receiver. For example, in GSM systems, stealing flags, which are time-multiplexed at predefined positions within a burst, are used to distinguish between

a FACCH burst and a TCH burst. By reading the stealing flags, the receiver determines the type of logical channels.

In systems that support multiple modulation schemes, demodulation of information communicated over control channels and traffic channels creates many complications. By introduction of link adaptation algorithms, adaptation of coding and/or modulation scheme becomes more frequent. The frequent link adaptations result in an increased signalling effort, causing degradation in communication quality. Furthermore, the control information communicated over FACCHs and voice or data communicated over TCHs must be demodulated without significant overhead in order to improve communication quality.

Therefore, there exists a need for an efficient and simple method for demodulating information in systems that support multiple modulation schemes.

SUMMARY

The present invention that addresses this need is exemplified in a method of demodulating variously modulated information using an identical demodulator in systems that support multiple modulation schemes.

Briefly, according to the method of the invention, voice or data is communicated over a traffic channel using a first linear modulation scheme, such as 16QAM or 8PSK modulation schemes. The traffic channel has an associated control channel that uses a second linear modulation scheme for communicating associated control information. In an exemplary embodiment, the second linear modulation scheme is a QPSK modulation scheme. The second linear modulation scheme, which has a lower level of modulation relative to the first modulation scheme, uses a reduced signal set of the first modulation scheme to communicate voice or data and control information. In this way, the present invention uses the same demodulator to demodulate signals modulated using the second linear modulation scheme as that used to demodulate signals modulated using the first linear modulation scheme.

According to some of the more detailed features of the present invention, the second modulation scheme uses the outer points of the modulation constellation of

the first modulation scheme. The first and second linear modulation schemes have the same symbol rate, the same pulse shaping, and the same burst format. Furthermore, the traffic channel and control channel use the same training sequences. In accordance with another aspect of the invention, the training sequences of the traffic channel and control channel are modulated using the second linear modulation scheme.

According to yet another aspect of the invention, voice or data are communicated using the first modulation scheme and in-band signalling information are communicated using the second modulation scheme. In this way, voice or data and in-band signalling information are demodulated using the same demodulation scheme, which corresponds to demodulation of signals modulated using the first modulation scheme. The in-band signalling information may include stealing flags that indicate whether a transmitted burst contains control information or voice and data information. Alternatively, the in-band signalling information may indicate at least one or more of a modulation type, a channel coding, or a speech coding used by a transmitted burst.

Other features and advantages of the present invention will become apparent from the following description of the preferred embodiment, taken in conjunction with the accompanying drawings, which illustrate, by way of example, the principles of the invention.

BRIEF DESCRIPTION OF THE DRAWINGS

FIG. 1 is a block diagram of a communication system which advantageously uses the present invention.

FIG. 2(a) and 2(b) are diagrams of modulation constellations of a 16QAM and QPSK modulation schemes, respectively.

FIG. 3 is a diagram of a subdivided RF channel that is used in the communication system of FIG. 1.

FIG. 4 is a diagram of a normal transmission burst transmitted on the RF channel of FIG. 2.

FIG. 5 is a block diagram of a mobile station used in the communication system of FIG. 1.

FIG. 6 is a block diagram of a radio base station used in the communication system of FIG. 1.

FIG. 7 is a block diagram of a radio transceiver used in the base station of FIG. 6.

5 FIG. 8 shows a diagram of the format of bits and symbols of a transmitted burst.

FIG. 9 shows a diagram of a mapping scheme used for demodulating the transmitted bursts of FIG. 8.

10 **DETAILED DESCRIPTION**

Referring to FIG. 1, a communication system 10 according to an exemplary embodiment of the present invention supports multiple modulation schemes. In an exemplary embodiment of the invention, the system 10 supports three modulation schemes: a first LLM (LLM1) scheme, a second LLM (LLM2) scheme, and a HLM scheme. In an exemplary embodiment, the first LLM (LLM1) scheme is a non-linear modulation scheme, such as GMSK modulation scheme used in GSM systems. A second LLM (LLM2) scheme is a linear modulation scheme, such as QPSK. Finally, the HLM modulation scheme is a higher level linear modulation schemes, for example, 16QAM or 8PSK scheme. LLM2 and HLM schemes have the same symbol rate that is different from the symbol rate of LLM1 scheme.

The mode of operation of GSM communication systems is described in European Telecommunication Standard Institute (ETSI) documents ETS 300 573, ETS 300 574 and ETS 300 578, which are hereby incorporated by reference. Therefore, the operation of the GSM system is described to the extent necessary for understanding of the present invention. Although, the present invention is described as embodied in a GSM system, those skilled in the art would appreciate that the present invention could be used in a wide variety of other digital communication systems, such as those based on PDC or D-AMPS standards and enhancements thereof. The present invention may also be used in CDMA or a hybrid of CDMA and TDMA communication systems.

The communication system 10 covers a geographical area that is subdivided into communication cells, which together provide communication coverage to a service area, for example, an entire city. Preferably, the communication cells are patterned according to a cell pattern that allows some of the spaced apart cells to use the same uplink and downlink RF channels. In this way, the cell pattern of the system 10 reduces the number of RF channels needed to cover the service area. The system 10 may also employ frequency hopping techniques, for example, to avoid "deadspots."

Referring to FIGs. 2(a) and 2(b), the signal sets in modulation constellations of 16QAM scheme and QPSK scheme are shown, respectively. The outer signal points of 16QAM scheme are shown by points A, B, C, and D, and the signal points of QPSK scheme are shown by points A', B', C', and D'. QPSK scheme can be viewed as having a reduced signal set relative to 16QAM scheme. If the symbol rates of QPSK and 16QAM schemes are the same, a 16QAM demodulator can demodulate the reduced signal set of QPSK modulation scheme by using exclusively the outer signal points A, B, C and D of 16QAM scheme. Consequently, the same demodulator can be used to demodulate signals that are modulated with QPSK and 16QAM schemes, if the same pulse shaping and burst format is used for both of these schemes. This arrangement significantly facilitates demodulation switching between QPSK and 16QAM schemes, for example, during link adaptation.

In one aspect, the present invention takes advantage of demodulation interchangeability with modulation schemes that have the same symbol rate, pulse shaping, burst format, and where one modulation schemes has a reduced signal set relative to the other, to effectively demodulate a first set of information that is modulated using a first linear modulation scheme and a second set of information that is modulated using a second linear modulation scheme that is different from the first linear modulation scheme. Preferably, the first linear modulation scheme has a higher level of modulation than the second linear modulation scheme. In this way, the present invention demodulates the first set of information and the second set of information using the same demodulation scheme that corresponds to demodulation of information modulated by the first modulation scheme.

In present GSM system, receivers treat GMSK modulation scheme as a linear modulation scheme. This means that a single demodulator can be used to demodulate GMSK and off-set QPSK modulated signals, as long as their symbol rates are the same. Similarly, a single demodulator can be used to demodulate GMSK and
5 higher level linearly modulated signals, as long as the signal points used by the demodulator during GMSK de-modulation are the reduced signal sets of a higher level modulation scheme and as long as the modulated signals have the same symbol rate.

The present invention communicates voice or data between a base station and a mobile station 12 over traffic channels. Voice and data are communicated over
10 the traffic channel using the first linear modulation scheme. For example, if possible, the first modulation scheme is preferably HLM scheme. Otherwise, voice or data are communicated using LLM2 scheme, which may be a QPSK modulation scheme. The present invention also communicates control information over associated or non-associated control channels. Preferably, the second modulation scheme of the associated
15 control channels and the first modulation scheme of the traffic channels have the same symbol rate, even though their modulation levels may be different, or they may be the same. The traffic channels have associated control channels for communicating associated control information between the base station and the mobile station 12. In an exemplary embodiment, the second modulation scheme of the associated control
20 channels is the second lower level modulation scheme LLM2, which is a QPSK modulation scheme.

The HLM and LLM2 modulation schemes use the same pulse shaping, symbol rate, and burst format. LLM2 scheme, however, uses a reduced signal set of the HLM scheme. As described above, this requirement allows for the use of an
25 identical demodulator at the receivers to demodulate outer signal points of 16QAM scheme and signal points of QPSK modulation scheme, which is used to communicate control information on associated control channels. As described later, in-band signalling information and training sequences are also communicated, preferably, using LLM2 scheme. Because LLM2 scheme uses a reduced signal set of HLM scheme,
30 HLM demodulator, in addition to HLM modulated signals, can also demodulate LLM2

modulated signals by detecting the outer signal points of HLM modulation constellations.

The system 10 is designed as a hierarchial network with multiple levels for managing calls. Using an allocated set of uplink and downlink RF links, mobile stations 12 operating within the system 10 participate in calls using allocated time slots. At a high hierarchal level, a group of Mobile Service Switching Centers (MSCs) 14 are responsible for the routing of calls from an originator to a destination. In particular, they are responsible for setup, control and termination of calls. One of the MSCs 14, known as the gateway MSC, handles communication with a Public Switched Telephone Network (PSTN) 18, or other public and private networks. The communication system 10 uses the present invention to provide for link adaptation, when mobile stations 12 within a cell move within coverage areas that support one or more of LLM1, LLM2, HLM schemes.

At a lower hierarchal level, each one of the MSCs 14 are connected to a group of base station controllers (BSCs) 16. The primary function of a BSC 16 is radio resource management. For example, based on reported received signal strength at the mobile stations 12, the BSC 16 determines whether to initiate a hand-over. Under the GSM standard, the BSC 16 communicates with a MSC 14 under a standard interface known as the A-interface, which is based on the Mobile Application Part of CCITT Signaling System No. 7.

At a still lower hierarchical level each one of the BSCs 16 controls, a group of base transceiver stations (BTSs) 20. Each BTS 20 includes a number of TRXs that use the uplink and downlink RF channels to serve a particular common geographical area. The BTSs 20 primarily provide the RF links for the transmission and reception of data bursts to and from the mobile stations 12 within their designated cell. In an exemplary embodiment, a number of BTSs 20 are incorporated into a radio base station (RBS) 22. The RBS 22 may be configured according to a family of RBS-2000 products, which is offered by Ericsson, the assignee of the present invention.

With reference to FIG. 3, an RF channel 26 (uplink or downlink) is divided into repetitive time frames 27 during which information are communicated.

Each frame 27 is further divided into time slots 28 that carry packets of information. Speech or data is transmitted during time slots designated as traffic channels (TCH_1, \dots, TCH_n). All signaling functions pertaining to call management in the system, including initiations, hand overs, and termination are handled via control information
5 transmitted over control channels.

In order to provide backward compatibility with GSM systems, the system 10 uses GMSK modulation scheme for communicating control information over the non-associated control channels. The mobile stations 12 use the slow associated control channels (SACCHs) to transmit associated control signals, such as an RX-LEV
10 signal, which corresponds to the received signal strength at the mobile station 12 and RX-QUAL signal, which is a measure of various levels of bit error rate at the mobile station 12, as defined by the GSM standard. Fast associated control channels (FACCHs) perform control functions, such as hand-overs, by stealing time slots allocated for TCHs. Fast signalling procedure is used to indicate whether a time slot
15 contains control or voice and/or data. In the present invention, the FACCHS and SACCHs may use LLM2 or HLM modulation schemes to communicate control information independent from the modulation scheme used for TCHs, if LLM2 and HLM are supported.

The BSC 16 instructs the RBS 22 based on measures of channel
20 characteristics of RF links between mobile stations 12 to the RBS 22. As described later in detail, the channel characteristics may be measured based on a number of parameters, including received signal strength at the mobile station 12, bit error rate at the mobile station 12, the multipath propagation property of the uplink RF channel, for example, time dispersion, or a combination of them.

25 The system 10 carries out the transmission of information during a time slot in a burst that contain a predefined number of coded bits. The GSM specification defines various types of bursts: normal burst (NB), frequency correction burst (FB), synchronization burst (SB), access burst (AB), and dummy burst. The normal burst, which has a duration of 576 μ s, is used both during the traffic and some control

signalling channels. The remaining bursts are primarily used for access and maintaining signal and frequency synchronization within the system.

As shown in FIG. 4, a normal burst 29 includes two separate data portions 30 during which digital data bits are communicated. The normal burst also includes tail and guard sections 31 and 32 as shown. Among other things, the guard section 32 is used to allow for up-ramping of the burst and for down-ramping of the bursts. The tail section 31 is used for demodulation purposes. All burst transmissions, except dummy burst transmissions, include training sequences. The training sequences are patterned with predefined autocorrelation characteristics. During demodulation process, the auto correlation characteristic of the training sequence helps in the synchronization of the received bit sequences over an RF channel. In the normal burst 29, a training sequence 33 is positioned in the middle of the burst between its data portions.

In order to compensate for propagation delays, the communication system uses a time alignment process by which the mobile stations 12 align their burst transmissions to arrive at the BTSs 20 in proper time relationship relative to other bursts transmissions. As described later, the mobile station 12 and the RBS 22 incorporate equalizers, which correlate received baseband bit sequences over the uplink or downlink RF channels with the training sequences, to provide correlator responses that correspond to the properties of multipath propagation. Based on the correlator responses, the receiver section of the BTS 20 generates a timing advance (TA) parameter, which corresponds to a propagation delay over the uplink RF channel. The mobile station 12 uses the TA parameter, which is transmitted from the RBS 22, for advancing or retarding its burst transmissions relative to a time reference.

With reference to FIG. 5, the block diagram of a mobile station 12 is shown. The mobile station 12 includes a receiver section 34 and a transmitter section 36, which are coupled to an antenna 38 through a duplexer 39. The antenna 38 is used for receiving and transmitting RF signals to and from the BTS 20 over allocated uplink and downlink RF channels. The receiver section 34 includes an RF receiver 40, which includes a local oscillator 41, a mixer 42, and selectivity filters 43 arranged in a well

known manner, for down-converting and demodulating received signals to a baseband level. The RF receiver 40, which is tuned by the local oscillator 41 to the downlink channel, also provides an RX-LEV signal on line 44 that corresponds to the received signal strength at the mobile station 12.

5 The RF receiver provides a baseband signal to a demodulator 46 that demodulates coded data bits representing the received speech, data and signaling information. Depending on the type of mobile station 12, the demodulator 46 can support one or more demodulation schemes corresponding to LLM1, LLM2, and HLM schemes. For example, the demodulator of a mobile station 12 subscribed to an
10 operator that supports LLM1 scheme may be capable of demodulating LLM1 modulated signals only. On the other hand, the demodulator of a mobile station 12 subscribed to an operator that supports all of the three modulation schemes is preferably capable of demodulating LLM1, LLM2, and HLM schemes.

 As described above, the demodulator 46 includes an equalizer (not
15 shown) that processes the coded bit pattern disposed on the training sequences, to provide correlator response that are used for predictive demodulation of the baseband signal. The equalizer uses the correlator responses to determine the most probable bit sequence for demodulation. As defined by the GSM specification, a channel
decoder/interleaver 50 also provides an RX-QUAL signal on line 48, which is a measure
20 of various levels of bit error rate at the mobile station 12. The mobile station 12 reports the RX-QUAL signal and the RX-LEV signal to the BSC 16 on a SACCH channel.

 Preferably, bursts modulated according to LLM2 and HLM scheme, i.e., 16QAM and QPSK schemes, use the same pulse shaping, symbol rate and burst format, and use the same training sequences. Both modulation schemes use the same signal
25 points to modulate the training sequence. For example, a 16 QAM modulator modulates the training sequence using outer signal points A, B, C, and D, (shown in FIG. 2(a)). Similarly, a QPSK modulated signal, which has a reduced signal set relative to 16QAM modulated signal, uses signal points A', B', C', and D' (shown in FIG. 2(b)) for
transmitting the training sequence. Although the training sequence used in bursts that
30 communicate control information is the same as the training sequence of bursts that

communicate of voice or data, in the present invention, the modulation scheme used for communicating the training sequence of a control channel is different from that of a traffic channel. Similarly, the in-band signalling information as well as the stealing flags are modulated using the outer signal point of the linear modulation constellation.

5 As described before, the mobile station 12 can use the same demodulator, i.e., a 16QAM demodulator, to demodulate the in-band signalling information, as well as the training sequences. This arrangement significantly facilitates decoding of both the training sequence and in-band signalling information of HLM and LLM2 modulated signals.

10 The channel decoder/de-interleaver 50 decodes and de-interleaves the modulated signal. The speech data bits are applied to a speech decoder 52 that decodes the speech pattern using one of a variety of speech decoding algorithms. After decoding, the speech decoder 52 applies an analog speech signal to a output device 53, e.g., a speaker, via an audio amplifier 54. The channel decoder 50 provides the
15 decoded data and signalling information to a microprocessor 56 for further processing, for example, displaying the data to a user.

The transmitter section 36 includes an input device 57, e.g., a microphone and/or keypad, for inputting voice or data information. According to a specified speech/data coding techniques, a speech coder 58 digitizes and codes the voice
20 signals according to a variety of supported speech coding schemes. A channel coder/interleaver 62 codes the uplink data according to a specified coding/interleaving algorithms, which improves error detection and correction at the BTS 12. The channel coder/interleaver 62 provides an uplink baseband signal to a modulator 64. The modulator 64 modulates the uplink baseband signal according to one or more of
25 supported modulation schemes. Similar to the demodulator 46, the modulator 64 of the mobile station 12 may support one or more of LLM1, LLM2, and HLM schemes.

The modulator 64 applies the coded signal to an up-converter 67, which receives a carrier signal from the up-converted signal local oscillator 41. An RF amplifier 65 amplifies the up-converted signal for transmission through the antenna 38.
30 A well known frequency synthesizer 66, under the control of the microprocessor 56,

supplies the operating frequency information to the local oscillator 41. The microprocessor 56 causes the mobile station 12 to transmit the RX-QUAL and RX-LEV parameters to the RBS 22 over the SACCH.

Referring to FIG. 6, an exemplary block diagram of the RBS 22 is shown to include a plurality of BTSs 20 that serve different geographical areas. Through a timing bus 72, the BTSs 20 are synchronized with each other. Voice and data information are provided to and from the RBS 22 through a traffic bus 74 that may be coupled, through the A-bis interface, to a public or private voice and data transmission line, such as a T1 line (not shown). Each BTS 20 includes TRXs 75 and 76 that communicate with the mobile station 12. As shown, two antennas designated as 24A and 24B are spaced accordingly to cover cells 77 and 78. The TRXs 76 are coupled to the antennas 24 through combiner/duplexers 80 that combine downlink transmission signals from the TRXs 76 and distribute the uplink received signals from the mobile station 12. The RBS 22 also includes a base station common function (BCF) block 68 that controls the operation and maintenance of the RBS 22.

Referring to FIG. 7, a block diagram of a TRX 76 is shown. The TRX 76 includes a transmitter section 86, a receiver section 87, a baseband processor 88 and a TRX controller 90. Through a corresponding antenna 24 (shown in FIG. 6), the receiver section 87 receives uplink signals from the mobile station 12. A down-conversion block 91 down-converts the received signal. After down-converting the received signals, the receiver section 87 samples its phase and magnitude, via a sampler block 92, to provide received bit sequence to the baseband processor 88. An RSSI estimator 94 provides an RSSI signal on line 95, which is a measure of the received signal strength. The RSSI estimator 94 may also measure noise disturbance levels during idle channels. The TRX controller 90, which is coupled to the traffic bus 74, processes the commands received from the BSC 16 and transmits TRX related information, such as various TRX measurements, to the BSC 16. Under this arrangement, the TRX 76 periodically reports the RSSI signal and noise disturbance levels to the BSC 16.

The baseband processor 88 includes a demodulator 96 that receives uplink baseband data from the receiver section 87. The demodulator 96 generates correlator responses that are processed in a well known manner to retrieve the uplink baseband data. The demodulator 96 may support demodulation of signals that are modulated
5 using one or more of LLM1, LLM2 or HLM schemes. The uplink baseband data is applied to a channel decoder 97 that decodes the baseband signal according to one or more supported channel decoding scheme. The channel decoder 97 places the decoded baseband signal on the traffic bus 78, for further processing by the BSC 16.

When transmitting downlink baseband data, the baseband processor 88
10 receives properly coded data or digitized speech information from the BSC 16 over the traffic bus 74 and applies them to a channel coder 102 that codes and inter-leaves speech and data according to one or more of supported channel coding schemes. The transmitter section includes a modulator 104 that modulates the supplied data bits according to one or more of LLM1, LLM2, and HLM schemes. The modulator 104
15 provides downlink baseband signals to an up-conversion block 106 for up-conversion. A power amplifier 108 amplifies the up-converted signal for transmission through a corresponding antenna.

In an exemplary operation, the system 10 establishes a call between a mobile station 12 and a RBS 20 using LLM1 on the SDCCH. Afterward, the mobile
20 station 12 remains in an idle mode, while monitoring PCHs for paging signals directed to it. The system 10, for example, uses one or a combination of the RX-QUAL, RX-LEV, or TA parameters, which are measures of channel characteristic of the RF link, to decide whether a inter-cell hand-over, an intra-cell hand-over, or a link adaptation procedure should be initiated or not. The initiation of a intra-cell link adaptation
25 procedure within coverage areas that support LLM1, LLM2, and HLM schemes is based on the channel characteristic of the RF link as well. The BSC 16 compares the channel characteristic parameter to corresponding thresholds to determine whether to perform a link adaptation, or an inter-cell or intra-cell hand-over.

When a call is requested, TCHs are allocated based on the capabilities of
30 both the mobile station 12 and BTS 20 to use LLM2 and HLM schemes. When only

LLM1 is supported, TCHs use LLM1. If the system 10, including the mobile station 12, can support LLM2 or HLM schemes, allocated TCHs use LLM2 or HLM schemes. If link quality is sufficient for HLM scheme, the system 10 uses HLM scheme for communicating over the allocated TCHs. Otherwise, the system 10 uses LLM2 scheme.

5 After hand-over is completed, a link algorithm for switching modulation within a cell continues. A concurrently filed patent application titled "**A LINK ADAPTATION METHOD FOR LINKS USING MODULATION SCHEMES THAT HAVE DIFFERENT SYMBOL RATES**", which is hereby incorporated by reference, discloses a link adaptation procedure that may preferably be used to perform link
10 adaptation in system 10.

While a call is ongoing, voice or data are communicated over the traffic channels using HLM scheme, when possible. If the BTS 20 detects a hand-over condition based on the channel characteristic of the RF link, according to one aspect of the invention, a method of communication between the mobile station 12 and BTS 20
15 initiates a hand-over on an associated control channel using LLM2 scheme. After the hand-over is complete, the mobile station 12 and the BTS 20 resume communication over TCH using HLM scheme. In this way, the present invention provides an easy hand-over method because the hand-over commands over FACCHs are communicated using a reduced signal set of HLM scheme that is easily demodulated by the same
20 demodulator used for demodulating HLM modulated voice or data over TCHs.

For the sake of maintaining compatibility with existing systems, the number of bits in one FACCH block that should be transmitted must remain the same. When using a higher level modulation scheme, like 16QAM modulation scheme, a significantly higher maximum number of bits may be transmitted. Using the greater bit
25 rate provided by 16QAM modulation scheme, a greater number of redundancy bits can be used to increase communication reliability of control information.

According to another aspect of the invention, the system 10 uses LLM2 to transmit control information over FACCH, independent of the modulation scheme used on TCHs, which may be one of LLM2 or HLM schemes. LLM2 scheme, which
30 has a lower level of modulation relative to HLM scheme, uses reduced signal set of

HLM modulation scheme to communicate control information. For example, LLM2 scheme may be QPSK modulation scheme and HLM scheme may be a 16QAM modulation scheme. In this way, both QPSK modulated signals and 16QAM modulated signals can be demodulated using a 16QAM demodulator. Consequently, data reliability
5 over FACCHs is improved compared to TCHs by increased Euclidean distance between signal points of modulation, that is, QPSK modulation scheme compared to 16QAM scheme. By this approach, the reliability is improved compared to traffic channels. Therefore, decoding complexity in terms of MIPS and memory is not increased compared to TCH processing, although reliability is improved. In another embodiment,
10 the system 10 uses HLM scheme with very low-rate convolutional coding to transmit control information over FACCHs.

Furthermore, the system 10 uses stealing flags to indicate whether a transmitted burst contains voice and data or control information. The stealing flags contained in the transmitted burst may be transmitted using either QPSK or 16QAM
15 modulation schemes. In case they are transmitted using QPSK modulation scheme, no additional bit is transmitted for stealing flags over TCHs. The advantage of transmitting the stealing flags using, QPSK modulation scheme, i.e., LLM2 scheme, is that they can be demodulated and evaluated independent of the modulation applied to voice or data.

Generally, SACCHs are transmitted on the same carrier as TCHs. The
20 position of SACCHs is well defined so that the receiver is able to demodulate SACCH bursts. In yet another aspect of the invention, LLM2 scheme is used for transmissions over SACCHs. In this way, the demodulation process is simplified, because the symbol rates of LLM2 and HLM are the same. The present invention may also use LLM2 scheme for SDCCHs and other control channel, such as PCHs and AGCH, in the same
25 way as that used for the SACCHs.

As described above, in-band signalling procedure places control signals in each burst, i.e., time-slot for TDMA systems, on pre-defined positions. According to another aspect of the present invention, in-band signaling is used to indicate at least one or more of a modulation type, a channel coding, and/or a speech coding used for a
30 transmitted burst. The present invention reserves a number of bits (or symbols), similar

to the stealing flags, as in-band signaling information to indicate which modulation scheme or channel coding scheme or speech coding is used in the transmitted burst. The reserved symbols or bits have a predefined location within the burst. In order to use the same demodulation scheme as that used to demodulate LLM2 or HLM modulated voice or data, the reserved bits or symbols are preferably modulated using LLM2 scheme. In this way, the receiver can demodulate and evaluate the in-band signaling information independent of the modulation scheme used for voice or data by using identical demodulation schemes. Therefore, the present invention may modulate in-band signalling information and voice or data using separate modulation schemes, but it demodulates them using the same demodulation scheme.

Referring to Fig. 8, a frame containing bits and symbols within a burst is depicted. Each 16QAM symbol comprises four bits. For transmission of data symbols, all four bits contain information that are estimated at receivers. For symbols that are used for in-band signaling, only two bits, bits 1 and 2, bear signalling information, the other two bits, bits 3 and 4, are set to zero. According to in-band signalling method of the invention, only the four outer signal points (in the corners of the 16QAM constellation) are used.

Referring to FIG. 9, a diagram of a mapping scheme used for demodulation of LLM2 and HLM modulated symbols is shown. As shown in FIG. 9, all four outer signal points have the bit pattern 'xy00', where x and y are equivalent to the bits 0 and 1 of the symbol used for in-band signaling. In this way, in-band signaling is efficiently used for transmission of fast control information, for example, to indicate the used modulation scheme.

System 10 transmits HLM and LLM modulated symbols next to each other. In many mobile radio systems, inter-symbol interference is handled by the equalizer in the receiver. Many equalizers use a-priori information on the used signal set. Such equalizers are for example based on maximum-likelihood sequence estimation, decision-feedback sequence estimation, etc. An embodiment of the invention is to apply an equalizer that assumes the signal set of HLM scheme for equalization of the whole

burst, although LLM2 symbols may be transmitted in this burst as well. The benefit of this procedure is that in-band signaling information can be evaluated after equalization.

From the foregoing it would be appreciated that the present invention significantly facilitates demodulation of information in a system that supports multiple modulation schemes, by reducing the overhead associated with demodulation of control
5 information and in-band signalling information. The present invention uses demodulation capability of a demodulator for higher level modulation to demodulate lower level modulated signals that have a reduced signal set. In this way, the present invention improves communication quality of systems that support multiple modulation
10 schemes.

Although the invention has been described in detail with reference only to a preferred embodiment, those skilled in the art will appreciate that various modifications can be made without departing from the invention. Accordingly, the invention is defined only by the following claims which are intended to embrace all
15 equivalents thereof.

What is claimed is:

1. A method for communicating information, comprising the step of:
modulating a first set of information using a first linear modulation
5 scheme;
modulating a second set of information using a second linear
modulation scheme, wherein the first and second linear modulation schemes have the
same symbol rate, and wherein the second modulation scheme uses a reduced signal set
of the first linear modulation scheme; and
10 demodulating the first set of information and the second set of
information using the same demodulation scheme.
2. The method of claim 1, wherein the same demodulation scheme
corresponds to demodulation of signals modulated using the first modulation scheme.
15
3. The method of claim 2, wherein the first linear modulation scheme has
a higher level of modulation than the second modulation scheme.
4. The method of claim 3, wherein the second linear modulation scheme
20 uses the outer signal points in modulation constellation of the first linear modulation
scheme to communicate the second set of information.
5. The method of claim 4 further including the steps of communicating
the first set of information over a traffic channel and the second set of information over
25 a control channel.
6. The method of claim 5, wherein the control channel is an associated
control channel.

7. The method of claim 6, wherein the associated control channel is a fast associated control channel.

8. The method of claim 6, wherein the associated control channel is a
5 slow associated control channel.

9. The method of claim 3, wherein the second set of information is in-band signalling information.

10. The method of claim 9, wherein the in-band signalling information
10 correspond to at least one of a modulation type, a channel coding or a speech coding.

11. The method of claim 3, wherein the second set of information
includes stealing flags that indicate whether a transmitted burst contains control
15 information or voice and data information.

12. The method of claim 3, wherein the second set of information is a training sequence.

13. The method of claim 3, wherein the first linear modulation scheme
20 and the second linear modulation scheme use the same pulse shaping.

14. The method of claim 3, wherein the first and second linear
modulation schemes use the same burst format.

15. The method in claim 3, wherein the first modulation scheme is a
25 QAM modulation scheme and the second modulation scheme is a QPSK modulation scheme.

16. The method in claim 3, wherein the first modulation scheme is a 8PSK modulation scheme and the second modulation scheme is a QPSK modulation scheme.

5 17. A method for communicating information between a base station and a mobile station comprising:

communicating voice or data using a first linear modulation scheme;

communicating in-band signalling information using a second
10 linear modulation scheme; and

demodulating the voice or data and the in-band signalling information using the same demodulation scheme.

18. The method of claim 17, wherein the first and second linear
15 modulation schemes have the same symbol rate, and wherein the second modulation scheme uses a reduced signal set of the first linear modulation scheme.

19. The method of claim 18, wherein the same demodulation scheme corresponds to demodulation of signals modulated using the first linear modulation
20 scheme.

20. The method of claim 18, wherein the in-band signalling information includes stealing flags that indicate whether a transmitted burst contains control information or voice and data information.

25

21. The method of claim 18, wherein the in-band signalling information indicates at least one of a modulation type, a channel coding, or a speech coding used by a transmitted burst.

22. A method of communication between a base station and a mobile station comprising:

modulating voice or data over a traffic channel using a linear first modulation scheme;

5 modulating control information over a control channel that is associated with the traffic channel using a second linear modulation scheme; and

demodulating the voice or data and control information using the same demodulation scheme.

10 23. The method of claim 22, wherein the first and the second linear modulation schemes have the same symbol rate.

24. The method of claim 23, wherein the same demodulation scheme corresponds to demodulation of signals modulated using the first linear modulation
15 scheme.

25. The method of claim 22, wherein the associated control channel is a fast associated control channel.

20 26. The method of claim 22, wherein the associated control channel is a slow associated control channel.

27. The method of claim 22, wherein the first linear modulation scheme has a higher level of modulation than the second linear modulation scheme.

25 28. The method of claim 27, wherein the second linear modulation scheme uses a reduced signal set of the first linear modulation scheme.

29. The method of claim 22, wherein the first and second linear
30 modulation schemes have the same level.

30. The method of claim 22, wherein the first and second linear modulation schemes use the same pulse shaping.

5 31. The method of claim 22, wherein the first and second linear modulation schemes use the same burst format.

32. The method in claim 22, wherein the first and second linear modulation schemes use the same training sequences.

10

33. The method in claim 22, wherein the first linear modulation scheme is a QAM modulation scheme and the second linear modulation scheme is a QPSK modulation scheme.

15

34. A demodulator comprising:

means for demodulating a first set of information that are modulated using a first linear modulation scheme; and

20 means for demodulating a second set of information that are modulated using a second linear modulation scheme, wherein the first and second linear modulation schemes have the same symbol rate, and wherein the second modulation scheme uses a reduced signal set of the first linear modulation scheme; and wherein the first set of information and the second set of information are demodulated using an identical demodulation scheme.

25

35. A method for communicating information, comprising the step of: modulating a first set of information using a first modulation scheme;

modulating a second set of information using a second modulation scheme, wherein the first and second modulation schemes have the same symbol rate,

and wherein the second modulation scheme uses a reduced signal set of the first modulation scheme; and

demodulating the first set of information and the second set of information using the same demodulation scheme.

5

36. The method of claim 35, wherein the first modulation scheme is a linear modulation scheme and the second modulation scheme is a non-linear modulation scheme.

10

37. The method of claim 36, wherein the non-linear modulation scheme is GMSK modulation scheme and the linear modulation scheme is a high level modulation scheme.

15

38. The method of claim 37, wherein the linear modulation scheme is 8PSK modulation scheme.

39. The method of claim 35 wherein the first modulation scheme has a higher level of modulation than the second modulation scheme.

20

40. The method of claim 35 further including the steps of communicating the first set of information over a traffic channel and the second set of information over a control channel.

25

41. The method of claim 39 wherein the control channel is an associated control channel.

42. The method of claim 38 wherein the second set of information is in-band signalling information.

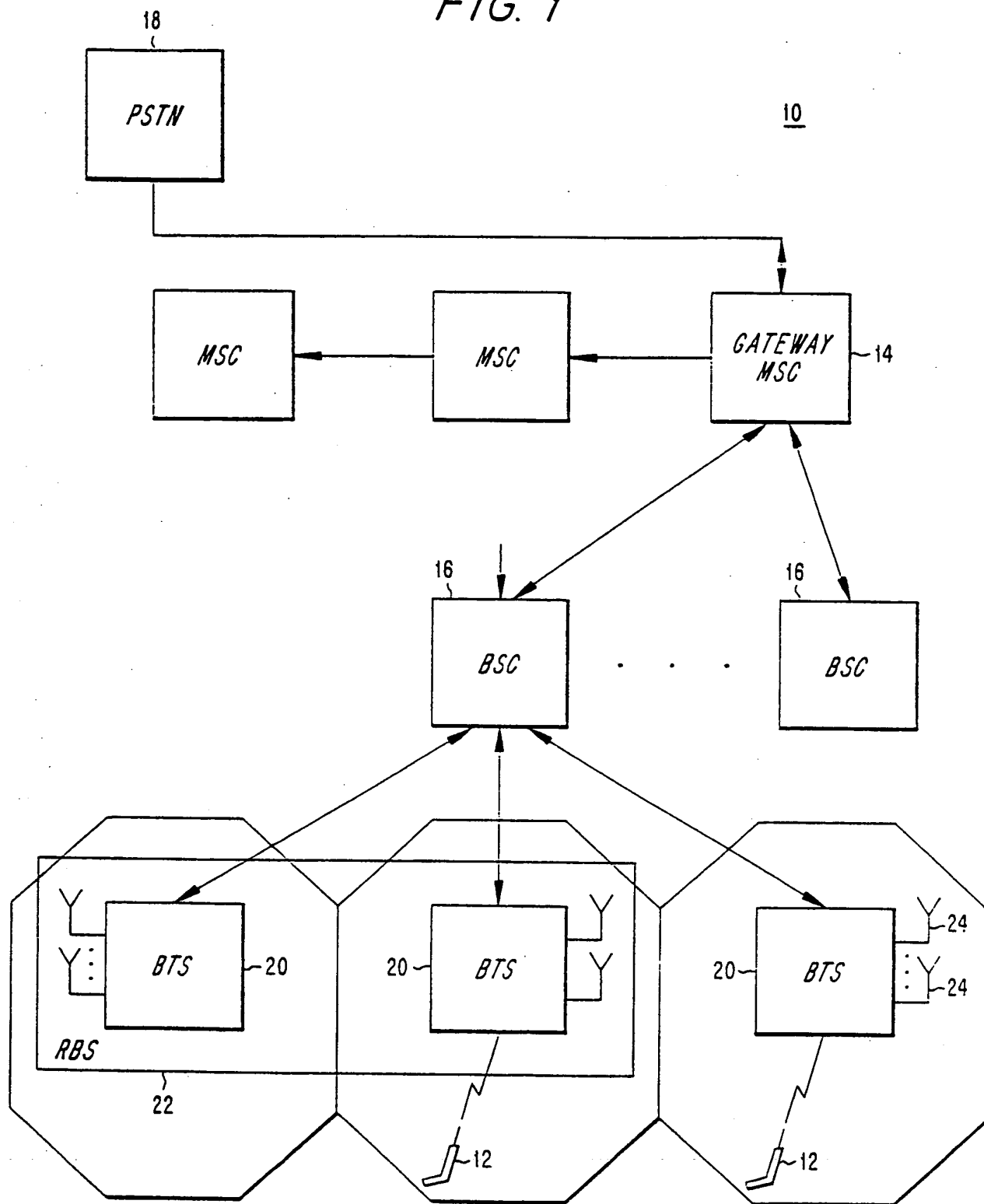
43. The method of claim 41, wherein the in-band signalling information correspond to at least one of a modulation type, a channel coding or a speech coding used for a transmitted burst.

5 44. The method of claim 35 wherein the second set of information includes stealing flags that indicate whether a transmitted burst contains control information or voice and data information.

 45. The method of claim 35 wherein the second set of information is a
10 training sequence.

1/6

FIG. 1



2/6

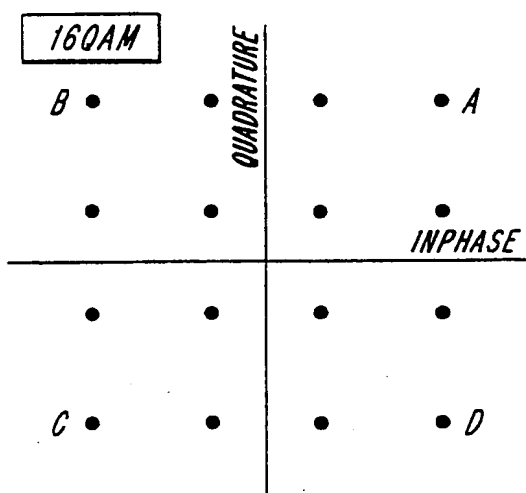


FIG. 2(a)

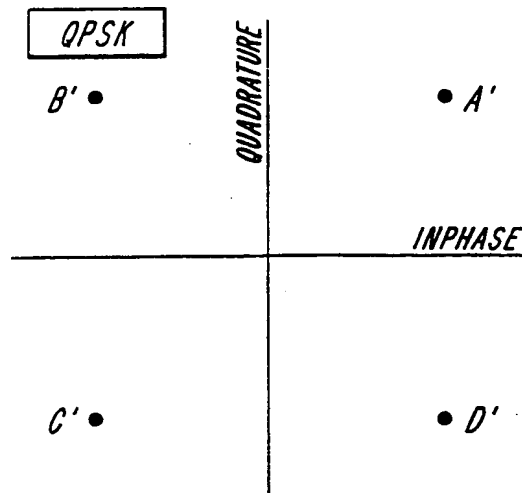


FIG. 2(b)

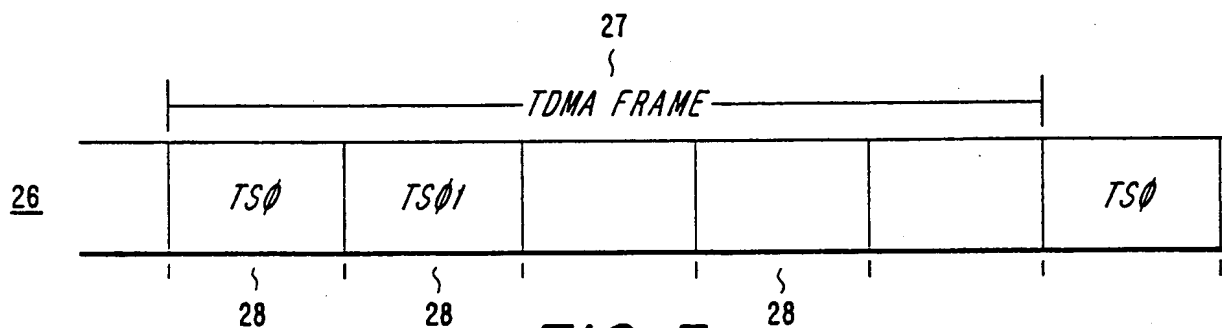


FIG. 3

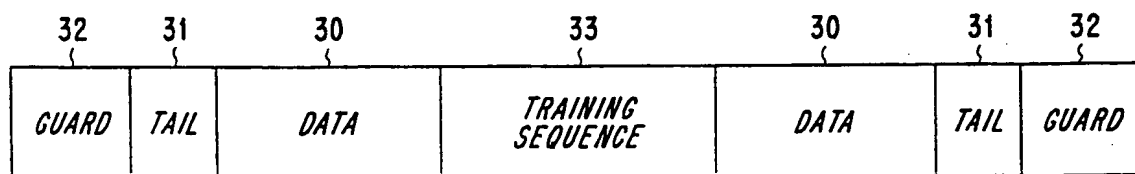


FIG. 4

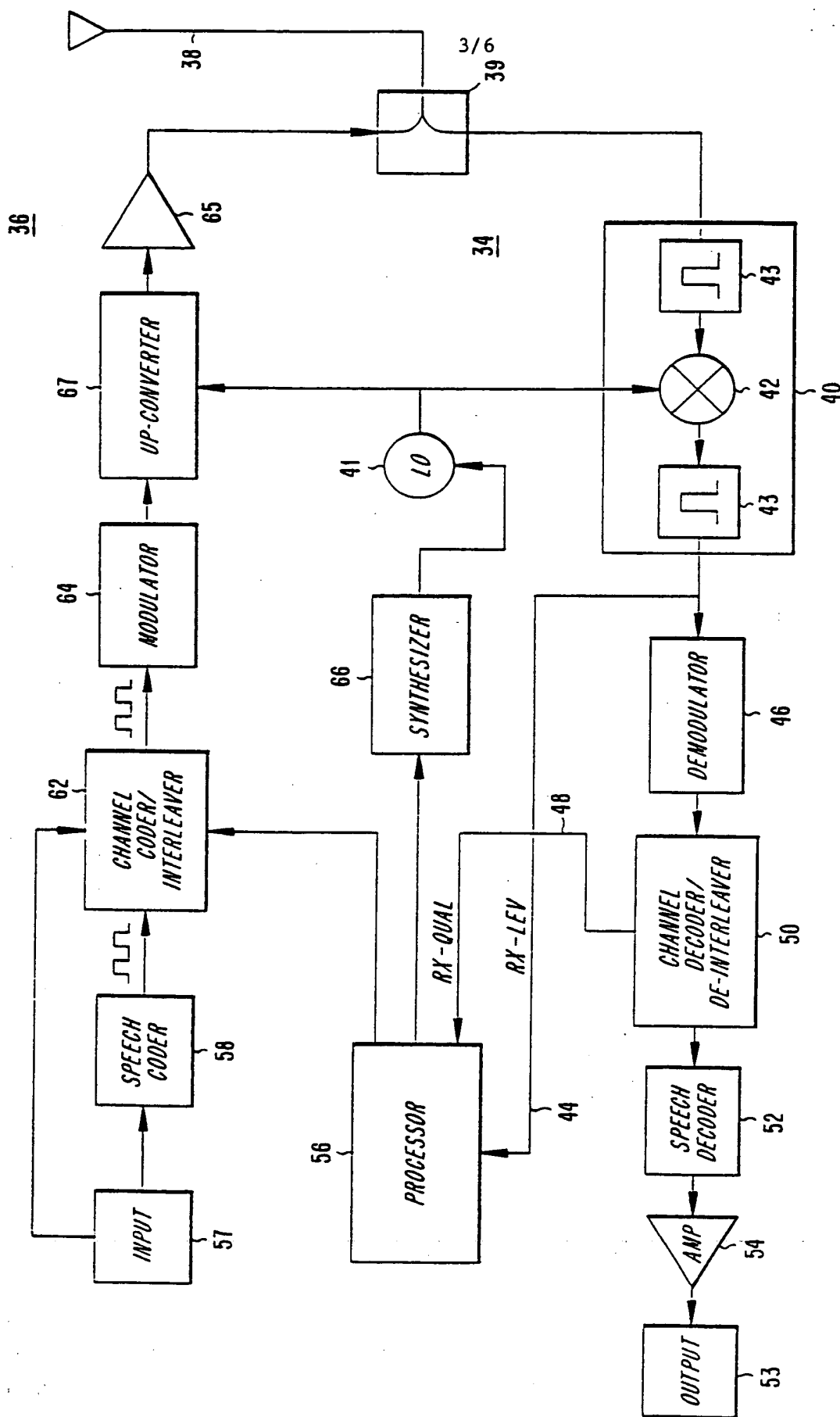


FIG. 5

4/6

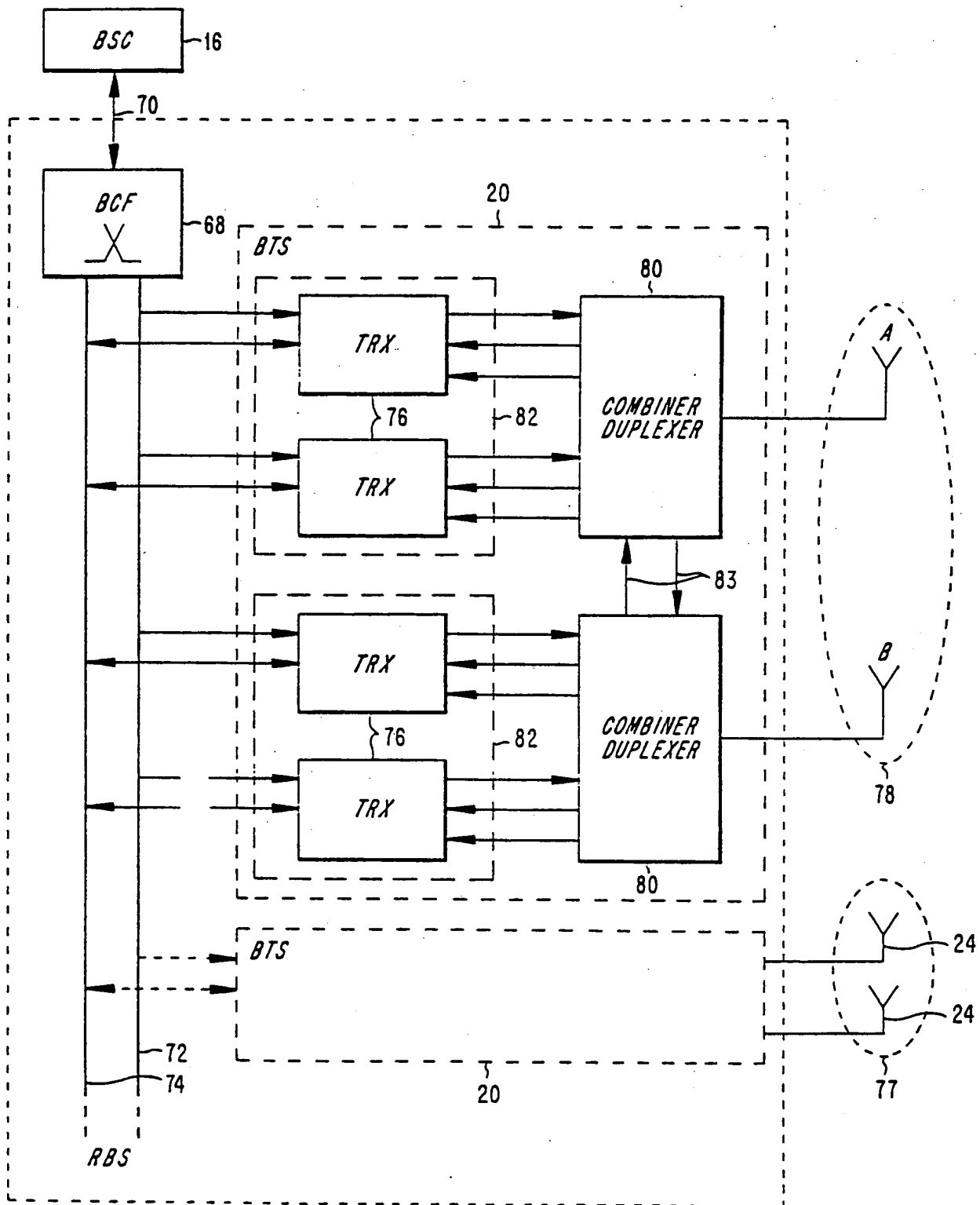


FIG. 6

5/6

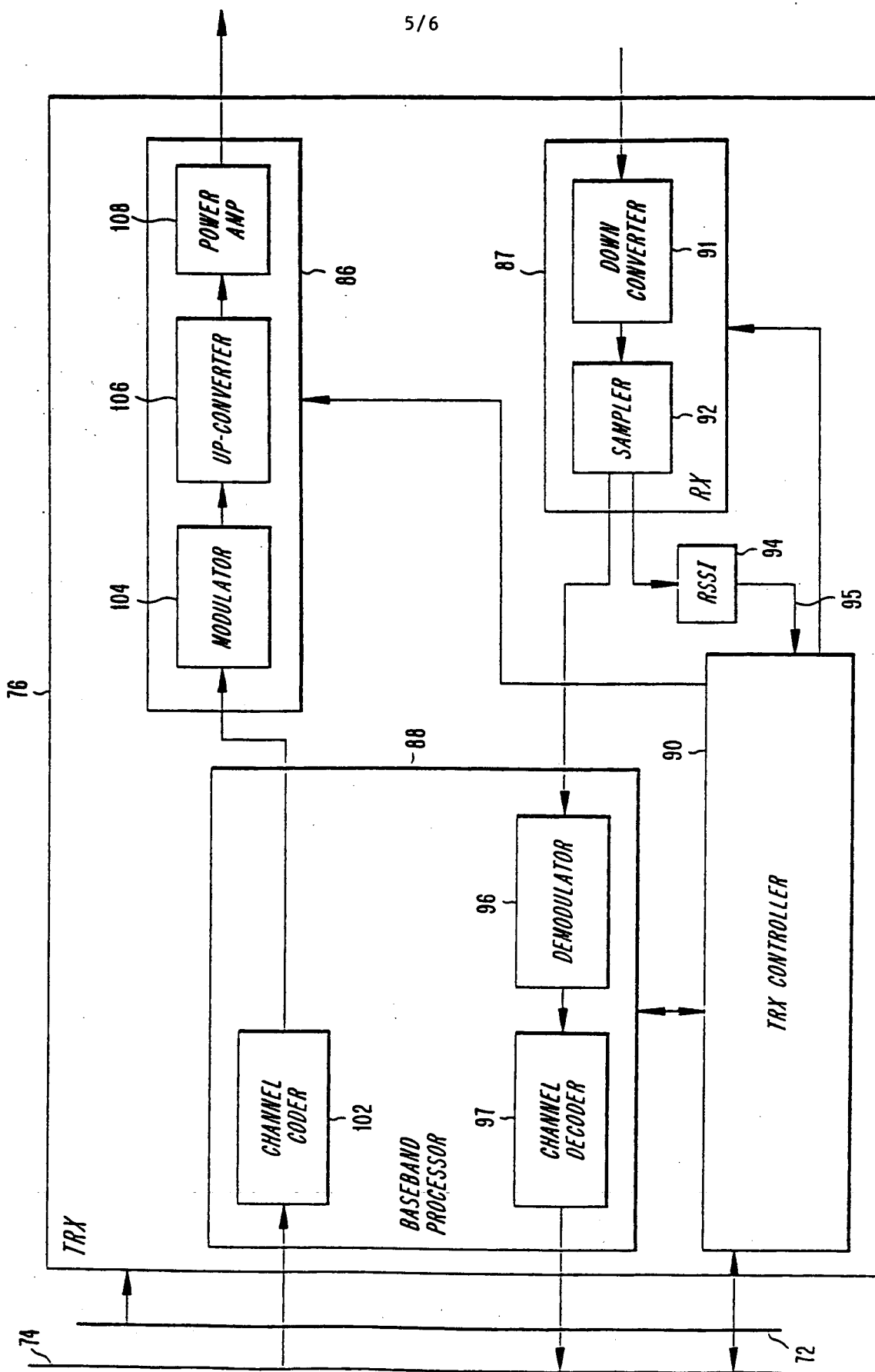


FIG. 7

6/6

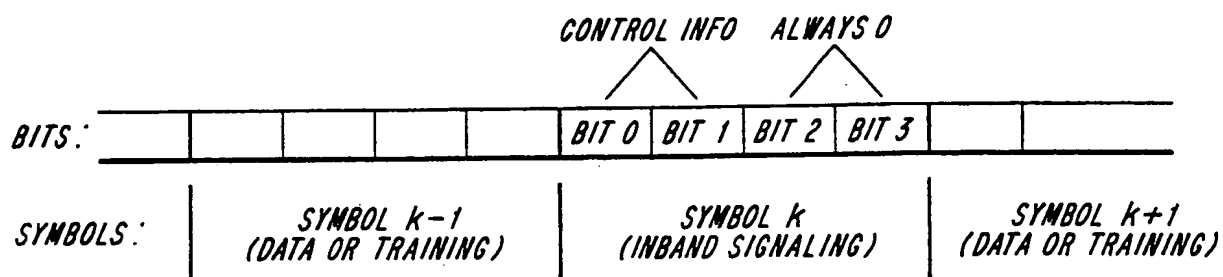


FIG. 8

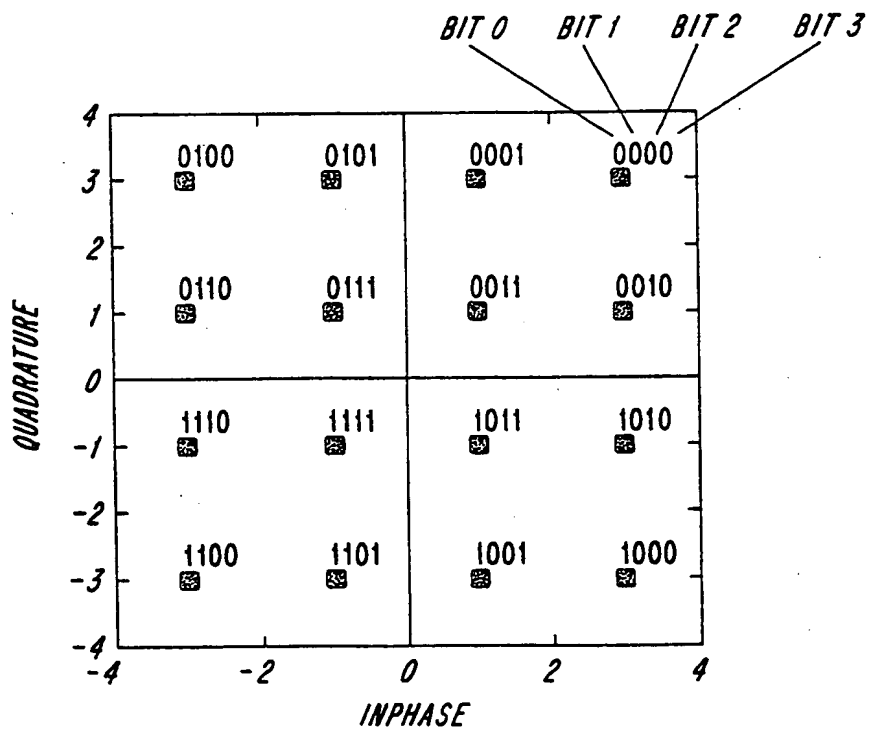


FIG. 9

INTERNATIONAL SEARCH REPORT

International Application No
PCT/SE 98/01540

A. CLASSIFICATION OF SUBJECT MATTER
IPC 6 H04B7/26 H04L27/34

According to International Patent Classification (IPC) or to both national classification and IPC

B. FIELDS SEARCHED

Minimum documentation searched (classification system followed by classification symbols)
IPC 6 H04B H04L

Documentation searched other than minimum documentation to the extent that such documents are included in the fields searched

Electronic data base consulted during the international search (name of data base and, where practical, search terms used)

C. DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	ITU-T Recommendation V.32 Data Communication over the Telephone Network March 1993 - Helsinki XP002092724 see page 1, paragraph 1 - page 3, paragraph 2.4.2 see figure 1 see table 3 --- -/--	1-5, 15-19, 22-24, 27, 28, 33, 35-39

☒ Further documents are listed in the continuation of box C.

☒ Patent family members are listed in annex.

Special categories of cited documents:

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"Z" document member of the same patent family

Date of the actual completion of the international search

9 February 1999

Date of mailing of the international search report

18/02/1999

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Lustrini, D

INTERNATIONAL SEARCH REPORT

International Application No

PCT/SE 98/01540

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
A	<p>ITU-T Recommendation V.34 Data Communication over the Telephone Network September 1994 - Geneva XP002092725 see page 1, paragraph 1 - page 3, paragraph 3 see page 11, paragraph 8.3 see page 13, paragraph 9.1 see page 26, paragraph 10.1.2.3 see figure 5</p>	<p>1,3,5, 17,22, 27,35-40</p>
A	<p>US 5 535 215 A (HIEATT III WILLIAM R) 9 July 1996</p> <p>see abstract see column 1, line 23 - line 26 see column 2, line 63 - line 65 see column 3, line 7 - line 21 see column 5, line 53 - line 58 see column 6, line 14 - line 24 see column 8, line 10 - line 25</p>	<p>1,5,17, 22, 34-36,40</p>
A	<p>JACOBSMEYER J M: "AN ADAPTIVE MODULATION SCHEME FOR BANDWIDTH-LIMITED METEOR-BURST CHANNELS" 21ST. CENTURY MILITARY COMMUNICATIONS - WHAT'S POSSIBLE ?, SAN DIEGO, OCT. 23 - 26, 1988, vol. 3, 23 October 1988, pages 933-937, XP000012361 INSTITUTE OF ELECTRICAL AND ELECTRONICS ENGINEERS see abstract see page 936, left-hand column see figure 4</p>	<p>1-4, 15-19, 22,27, 28, 33-37,39</p>
A	<p>WO 92 22162 A (BRITISH TELECOMM) 10 December 1992</p> <p>see abstract see page 2, line 1 - line 17 see page 5, line 15 - line 20 see page 6 see page 11, line 24 - line 39 see figure 2</p>	<p>1-4,10, 15-18, 20-22, 27,28, 33-39, 43,44</p>

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INTERNATIONAL SEARCH REPORT

Int. Jonal Application No
PCT/SE 98/01540

C.(Continuation) DOCUMENTS CONSIDERED TO BE RELEVANT

Category ²	Citation of document, with indication, where appropriate, of the relevant passages	Relevant to claim No.
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A	<p>US 4 866 395 A (HOSTETTER G ROBERT) 12 September 1989 see abstract see column 2, line 67 - column 3, line 7 see figures 1A-1F see figures 3,4A-4C see figure 9</p>	<p>34</p>

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information on patent family members

International Application No

PCT/SE 98/01540

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